



# Development Of An AI – Assisted Hysteresis Current Controlled 3-Level Diode Clamped Multilevel Inverter For Three Phase Bidirectional G2V/V2G EV DC Fast Charging Architecture

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**Abstract:** The rapid growth of electric vehicles has increased the demand for efficient DC fast-charging systems with bidirectional power transfer capability. This study presents a 3-Level Diode Clamped Multilevel Inverter (DCMLI) integrated with AI Assisted Hysteresis Band Current Control (AIHBCC) for Grid-to-Vehicle (G2V) and Vehicle-to-Grid (V2G) operations. The proposed system consists of a three-level NPC inverter, bidirectional DC-DC converter, DC-link capacitor, grid filter, and EV battery. HBCC ensures accurate current tracking, fast dynamic response, and improved power quality, while PLL and PI controllers maintain grid synchronization and DC-link voltage stability. MATLAB/Simulink results demonstrate reduced harmonic distortion, stable DC-link voltage, effective bidirectional energy flow, and enhanced charging/discharging performance. The proposed topology provides an efficient and reliable solution for future smart-grid-integrated EV fast-charging applications.

**Keywords :-** DC Fast Charging, V2G Operation, G2V Operation, NPC Multilevel Inverter, Buck -Boost Converter, Hysteresis Current Control.

## 1.Introduction

The fast growth in the field of electric vehicles has strengthened into an essential solution for emissions of greenhouse gases, environmental issues with standard ignition engine vehicles and the exhaustion of fossil fuels. Compared to conventional transportation systems, EV technology promotes increased energy efficiency and a lower carbon impact. However, widespread EV adoption poses serious problems for power networks, especially regarding variations in load demand, grid stability, and distribution infrastructure limitations.

V2G and G2V technologies have potential to turn EVs into supplied energy storage devices i.e. batteries that can exchange electricity with the grid in both directions, according to recent research. While stored energy is returned to the grid during times of peak demand, EV batteries in G2V mode absorb excess energy during off-peak hours. Supplementary services involving reactive power support, frequency management, peak cut off and gap filling are made possible by this idea [1], [2]. V2G technology, in comparison to traditional charging systems, boosts grid flexibility and facilitates the amalgamation of renewable energy.

Implementing V2G systems in micro-grid situations, where integration is relatively easier than in major utility grids, has been the subject of several studies [3]. Efficient energy storing systems are necessary to handle interruption in micro-grids by using non-conventional energy options like solar and wind power. Use of EV batteries function as dynamic storage resources when connected via intelligent charging infrastructure [4].

On board charger rating place limitation on conventional level -1 and level -2, AC charging approaches rendering them inappropriate for large capacity bidirectional energy exchange. level 3 off board DC fast charging topologies

have been suggested as a solution to these issues [5].

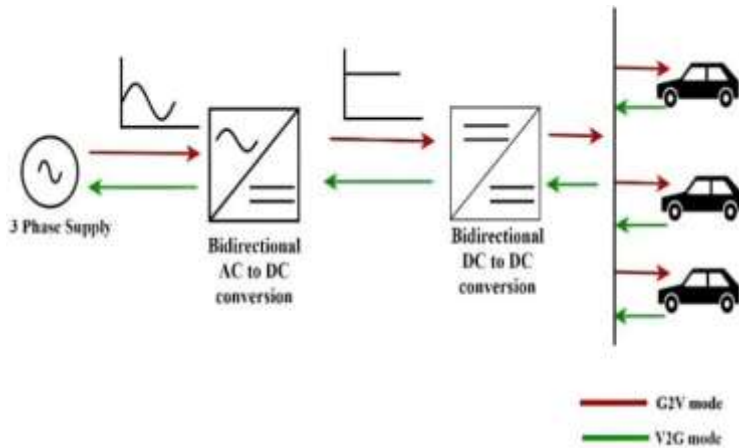
DC fast charging stations use grid connected inverters, LCL filters, and off board bidirectional DC -DC Converters to enable high power transfer while preserving acceptable harmonic performance. Cascaded vector regulator for grid connected inverters and continuous current control for battery chargers are examples of advanced control techniques which stabilized DC voltage and effective regulation of active as well as reactive power [6],[7]

Despite encouraging developments, increased EV penetration may give rise to harmonic distortion, increase distribution transformer loading, and impact power quality if improperly coordinated [8], [9]. Therefore, to determine if large-scale V2G deployment is technically feasible, modeling, simulation, and effect assessment studies are crucial. A detailed investigation of DC fast charging-based V2G-G2V topologies in micro-grid systems is provided in this review study. Power electronic setups, control strategies, harmonic reduction techniques, and grid integration issues identified in recent literature are all critically examined. The objective is to offer a comprehensive technical understanding of EV-based bidirectional energy systems and their function in the development of future smart grids [10].

## 2. DC fast charging station design with V2G-G2V operation

Fig. 1 shows the DC fast charger design for V2G-G2V structure in micro-grid via off-board charger. A Bidirectional AC to DC converter as a Diode Clamp Multilevel Inverter with transformer (step-up) and L-C-L filter via DC bus to utility grid. Below are described the main components of charging station [10].

**Fig. 1. G2V and V2G mode of operation of EV**



### 2.1 Configuration of Battery Charger

In DC fast charging Vehicle-to-Grid (V2G) systems, the charger of battery is a crucial port that enables bidirectional energy exchange between an EV battery and a micro-grid of DC bus. DC fast charging stations utilize off-board bidirectional DC-DC converters in contrast to traditional on-board chargers to achieve higher power capability, improved thermal performance, and enhanced control flexibility [4]. This arrangement is especially well suited for V2G implementation in micro-grid environments and high-power level-3 charging applications [11].

The bidirectional DC-DC converter which is usually made up of two actively controlled semiconductor switches like IGBTs or MOSFETs, by using DC-link capacitor, anti-parallel diodes, and an inductor, operates in two fundamental modes: buck mode during charging (G2V) and boost mode during discharging (V2G) [12], [13]. To control power flow and avoid short circuit situations, corresponding switching signals are employed. The working mode and power transfer magnitude are determined by the current reference that the controller generates [10]. Fig. 2 indicates the converter arrangement. It is made by using double IGBT's or MOSFET switches that always helps in controlling by complementary signals.

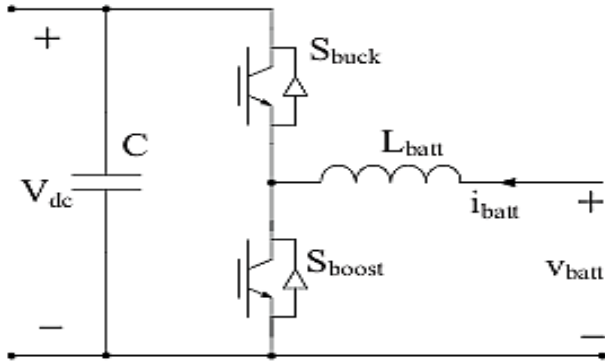
#### 2.1.1 Operation of Buck Mode (Grid to Vehicle - Charging Mode)

Fig. 2 shows a converter that functions as buck converter, reduce the input DC voltage  $V_{dc}$  to battery charging voltage  $V_{batt}$  when upper buck switch  $S_{buck}$  in operation mode. The current passes to battery via inductor and upper

switch while it is in the on state. In this charging process, power is moved from grid to vehicle (G2V). In the closed switch condition, the current returns through the diode connected at lower end and through inductor. Thus, completing the circuit. The upper switch's duty ratio decides the voltage of the battery[14].

$$V_{batt} = V_{DC} * D$$

**Fig. 2. Battery charger/Discharging Operations**



**2.1.2 Operation of Boost mode (Vehicle to grid discharging mode):**

When  $S_{Boost}$  is turned on, the converter operates as boost converter as illustrated in fig.2. Although lower switch  $S_{Boost}$  is operated, current can flow through the inductor, completing the path via the capacitor and diode of the  $S_{buck}$  switch connected in antiparallel. In this situation, battery enters in discharge mode and net power is directed from EV to grid (V2G). Under boost mode, output voltage will be equal if large capacity capacitors are used to force a continuous DC voltage [10]:

$$V_{DC} = \frac{V_{batt}}{1-D'}$$

Where  $D'$  represents the lower switch's duty cycle.

**2.2 Operational Significance in V2G Systems**

Intelligent control methods, which usually use PI-based constant current control schemes, regulate the shift between buck and boost modes [6]. Under various grid conditions, these controllers guarantee smooth bidirectional power transfer, control battery current, and preserve DC bus voltage steadiness.

Thus, in contemporary micro-grid systems, the bidirectional DC-DC buck boost converter is the essential component of a DC fast charging station, allowing for high-power, quick response, and effective V2G-G2V operation [15-20].

In order to enable high-power, quick reaction, and effective V2G-G2V operation in contemporary micro-grid systems, the bidirectional DC- DC buck boost converter is the essential component of fast charging station [10].

**2.3 Active Front End Converter as Multilevel inverter and LCL Filter connected to grid**

**2.3.1 Active Front End Converter as multilevel inverter connected in grid**

In DC fast charging, an A.F.E i.e. active front-end converter as Multilevel inverter to grid for AC -DC power conversion stage is widely used. The AFE converters uses as controlled semiconductor switch such as IGBT or MOSFET with advanced modulation technique to achieve bidirectional power flow, improve power quality and high efficiency.

In EV fast charging architecture, AFE converter is connected between grid and DC link of charger. The primary function of AFE Converter is three phase AC Grid voltage to regulate DC Voltage. which is supplied to DC-DC Buck – Boost Converter.

The AFC as a DCML converter also reduced harmonic content and improved power quality also. Those are essential specifications for contemporary power electronics converters connected to the grid.

The AFE converter generally consists of Multilevel Inverter with three phases implemented using twelve controlled switches arranged in a Diode Clamp NPC configuration.

This study proposes a novel bidirectional EV fast charging architecture based on a hysteresis current controlled

three-phase three-level diode-clamped NPC Multilevel inverter integrated with a bidirectional buck–boost converter. The proposed system addresses the limited application of hysteresis current control in EV fast charging and bidirectional energy exchange systems. The key contribution is the coordinated implementation of multilevel power conversion and hysteresis-based current regulation to achieve enhanced current tracking, lower total harmonic distortion, improved grid power quality, and efficient G2V/V2G operation. The proposed configuration offers a practical and computationally simple alternative to conventional PWM- and model-based control techniques while maintaining high dynamic performance under varying charging and discharging conditions.

An LCL filter is installed between the Front-end converter and grid. The filter reduces switching harmonics produced by the converter.

The major advantages of using an AFE Converter in EV fast charging stations include bidirectional operation for vehicle to grid and grid to vehicle application, reduced THD, power factor nearly one and improved dynamic performance. Due to these benefits, AFE converters are widely adopted in high power EV fast charging ranging from 10 KW to several hundred kilowatts [4].

### 2.3.2 Multilevel inverter

Multilevel inverters (MLIs) are widely used in medium- and high-power applications due to their ability to produce near-sinusoidal output voltages with reduced harmonic distortion. By increasing the number of voltage levels, MLIs achieve smoother output waveforms, lower switching losses, reduced dv/dt stress, and improved power quality compared to conventional two-level inverters. Additionally, they offer lower common-mode voltage, reduced current ripple, and efficient operation under both fundamental and high-frequency switching conditions [21].

Multi-level inverters have three major topologies that are Cascaded H-Bridge topology, Flying Capacitor topology and Neutral Point Clamped or Diode Clamped topology. The proposed system consists of Neutral point clamped Topology (Diode-Clamped Topology) [22]. The system is a Voltage Source Inverter because the input voltage level is kept constant. The basic concept of this topology is to use diodes and to generate multi level voltages through different phases. The 3-Level approach reduces the complexity of switching circuit. The system consists of three legs. Each leg consists of series connected four IGBT Switches. Two diodes are connected in parallel across each leg to avoid the turn-on of wrong pair of switches [23].

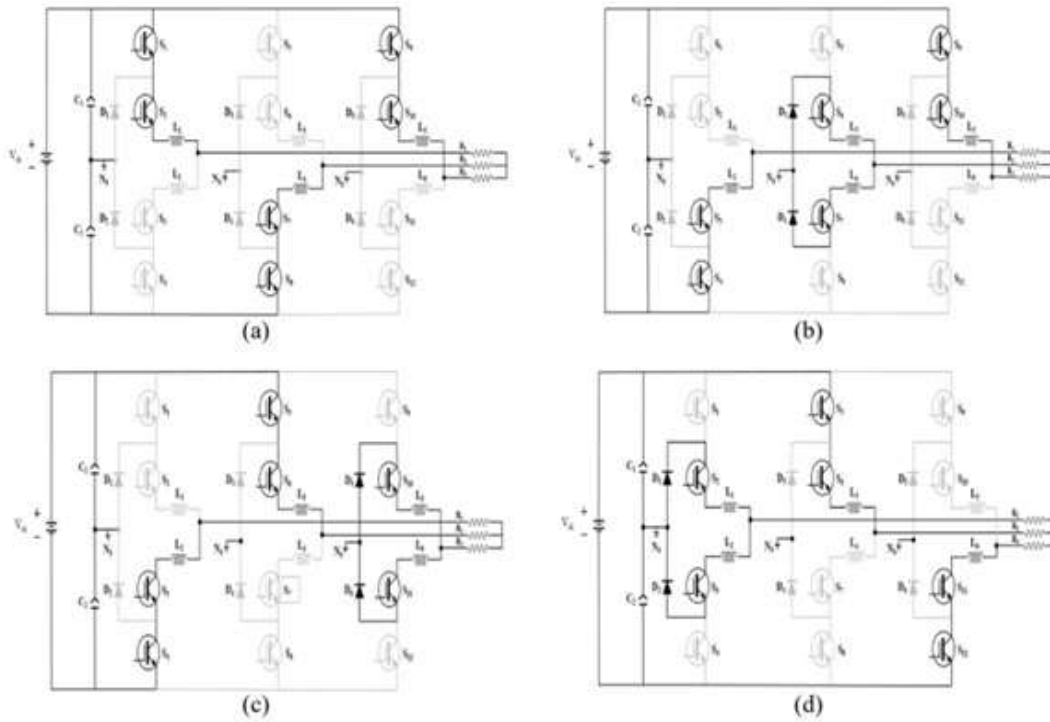
Input Voltage is  $V_{DC}$ . S1, S2, S3 and S4 are the switches in Leg-1. These switches can be configured in three switching states, S1-S2 ON and S3-S4 OFF leading to output  $\frac{+V_{DC}}{2}$ ; S2-S3 ON and S1-S4 OFF leading to output 0 as the diodes short circuit and no outputs is generated; S3-S4 ON and S1-S2 OFF leading to output  $-\frac{V_{DC}}{2}$ . S5, S6, S7, S8 and S9, S10, S11 and S12 respectively. Similarly Leg-2 and Leg-3 have 3 configurations each. combining all these switching sequences, we can get a total of 27 configurations [24].

### 2.3.3 Mode of Operation with reduced switches

The three-level Neutral Point Clamped (NPC) inverter consists of three legs, each containing four IGBT switches, resulting in a total of 27 possible switching states. Figure 3(a) illustrates one operating mode in which switches S1 and S2 are turned ON while S3 and S4 are OFF in the first leg, producing an output voltage of  $\frac{+V_{DC}}{2}$ . In the second leg, switches S7 and S8 are ON and S5 and S6 are OFF, generating an output voltage of  $-\frac{V_{DC}}{2}$ . Similarly, in the third leg, switches S9 and S10 are ON while S11 and S12 are OFF, producing an output voltage of  $\frac{+V_{DC}}{2}$ . The corresponding current paths flow through the activated switches, enabling the synthesis of the desired three-level output voltage waveform.

Similarly works shown in Figure 3(b) -3(d).

**Figure 3. Modes of Operation (a)S1-S2, S7-S8, S9-S10 are ON; (b)S3-S4, S6-S7, S9-S10 are ON;(c)S3-S4, S5-S6, S10-S11 are ON; (d) S2-S3, S5-S6, S11-S12 are ON**

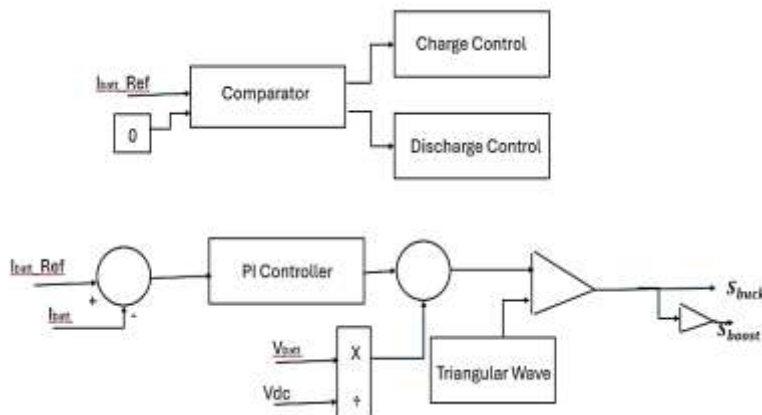


### 3. Control Strategies

#### 3.1 Constant Control strategies for control battery charger

Figure 3 illustrates a constant current control scheme [10] that relies on PI controllers to manage the battery's charge and discharge cycles. The system first determines the operating mode by checking the polarity of the reference current specifically, whether it is above or below zero. Once the appropriate mode is set, the system calculates the error between the reference current and actual measured current. This error signal is then processed by the PI controller for generating the necessary switching pulses to operate the circuit correctly [25-29].

Fig. 3. Battery charger constant current control strategy

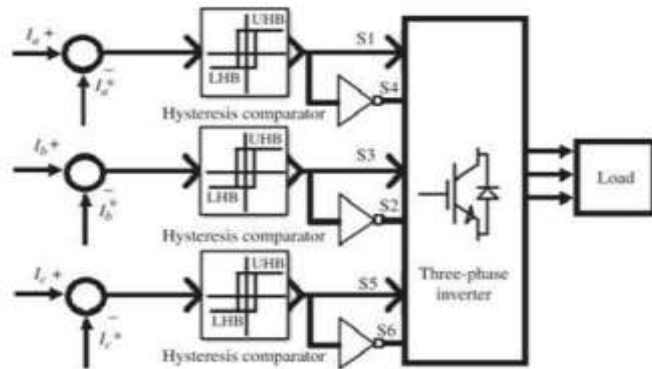


#### 3.2. Design of a Hysteresis Current Control.

The Hysteresis Current Controller (HCC) is employed to regulate the grid current by continuously comparing the reference current with the actual grid current. The resulting current error is processed through a hysteresis comparator defined by an Upper Hysteresis Bound (UHB) and a Lower Hysteresis Bound (LHB). Whenever the

current error exceeds the prescribed hysteresis limits, the controller generates appropriate switching signals to force the inverter current back within the hysteresis band. Consequently, the actual current accurately tracks the reference current, ensuring fast dynamic response and effective current regulation. The output of the hysteresis comparator is directly applied to the gate terminals of the inverter switches, thereby generating the required gating pulses for inverter operation. This control technique offers simple implementation, high tracking accuracy, improved harmonic performance, and rapid transient response, making it suitable for bidirectional EV charging and grid-connected power conversion applications.

**Figure.4 Hysteresis Current Controller**



The Hysteresis Current Controller (HCC) maintains the inverter output current within a predefined hysteresis band around the reference current. The controller continuously monitors the current error, which is the difference between the reference current and the actual grid current. When the current error reaches the Upper Hysteresis Bound (UHB), the inverter applies a negative voltage to the grid, causing the grid current to decrease. Conversely, when the current error reaches the Lower Hysteresis Bound (LHB), a positive inverter voltage is applied, resulting in an increase in grid current. This switching action ensures that the actual current closely follows the reference current trajectory[30].

In a three-phase grid-connected system, three independent hysteresis current controllers are implemented, one for each phase, to achieve accurate current tracking and balanced operation. The continuous increase and decrease of current within the hysteresis band enables rapid dynamic response and minimizes current tracking error. Furthermore, the reference current extraction algorithm generates synchronized current references, while the load-balancing strategy improves the distribution of power among phases and enhances real power injection into the utility grid. As a result, the HCC technique provides superior current regulation, improved power quality, reduced harmonic distortion, and efficient utilization of renewable energy sources in grid-connected applications [31].

### 3.3 Design of a AI-Assisted Hysteresis Current Control

An intelligent improvement of traditional hysteresis current control, AI-assisted hysteresis current control (AI-HCC) uses an artificial intelligence algorithm, such as an Artificial Neural Network (ANN), Fuzzy Logic Controller (FLC), or Adaptive Neuro-Fuzzy Inference System (ANFIS), to adaptively regulate the hysteresis band or switching strategy based on current operating conditions. The AI-assisted method continuously modifies the control parameters in accordance with current error, grid voltage, load fluctuations, and battery state of charge (SOC), in contrast to the traditional controller with a fixed hysteresis band [32],[33],[34].

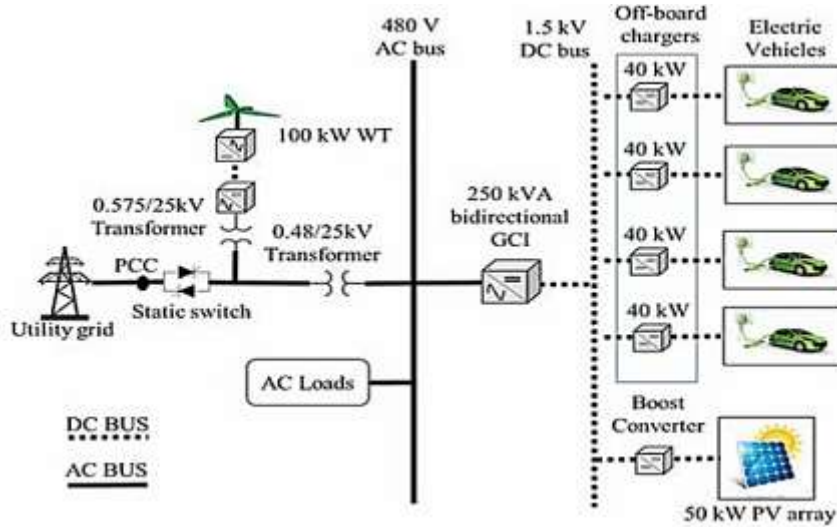
In bidirectional Grid-to-Vehicle (G2V) and Vehicle-to-Grid (V2G) electric vehicle DC fast-charging systems, this adaptive operation improves current tracking accuracy, minimizes current ripple and total harmonic distortion (THD), lowers switching losses, and increases the efficiency and dynamic performance of three-level diode-clamped multilevel inverters.

## 4. Micro – Grid Test System configuration

Configuration of Test system for the micro-grid, and dc fast charging station are illustrated in figure 5 below. A 100-kW wind turbine and a 50-kW solar PV system power the system. The four EV batteries are connected to a 1.5 kV

dc bus via off-board chargers at the charging station. This arrangement forms the EV battery storage system. The solar PV system is connected to a boost converter with maximum power point tracking control through DC bus. The microgrid is further connected to DFIG driven wind turbine.

Fig 5. Micro Grid Test System



## 5. Designing and Simulation Result

### 5.1 Buck-Boost Converter Design

(Bidirectional Battery Interface for V2G/G2V Operation)

- **System Specifications**

Battery Nominal Voltage  $V_{bat} = 360$  Volt

Output Current = 30 Amp

$V_{ripple} = 0.36$

$I_{ripple} = 3$  A

Switching Frequency  $f_{sw} = 10$  kHz

Inductor = 2 Mh

Capacitor C = 0.625  $\mu$ F

#### Design of Buck-Boost Inductor

##### Step -1 Calculation of a Duty Cycle.

For buck - boost converter,

$$D = \frac{V_o}{V_o + V_{in}} \text{ --- (1)}$$

$$D = \frac{360}{360 + 800}$$

$$D = 0.31$$

##### Step - 2 Calculation of a Inductor Current.

$$I_L = \frac{I_o}{1-D} \text{ --- (2)}$$

$$= \frac{30}{0.69}$$

$$= 43.4 \text{ Amp}$$

##### Step - 3 Calculation of a Ripple Current.

Ripple current is considered as 20% to 40% of a load current.

$$\Delta I_L = 0.3 * 43.4 = 13 \text{ Amp}$$

**Step - 4 Calculation of the Value of inductor.**

$$L = (V_{in} * D) / (F_s * \Delta I_L) \text{ --- (3)}$$

$$= \frac{800 * 0.31}{10000 * 13}$$

$$= 0.0019 \text{ H}$$

$$= 1.9 \text{ mH}$$

**Current rating,**

$$I_{peak} = I_L + \left(\frac{\Delta I_L}{2}\right) \text{ --- (4)}$$

$$= 43.4 + \left(\frac{13}{2}\right) = 49.9 \text{ Amp}$$

So, consider value of inductor L = 20 mH and 50 Amp

**Design of Buck-Boost capacitor,**

Output capacitor is selected to limit output voltage ripple.

Assume voltage ripple is consider as  $\Delta V_o = 1.5 \text{ Volt}$

**Output Capacitor is given by ,**

$$C = (I_o * D) / (F_{sw} * \Delta V_o) \text{ --- (5)}$$

$$= \frac{30 * 0.31}{(10000 * 1.5)}$$

$$= 0.00000062 \text{ F} = 0.62 \mu\text{F}$$

So, consider as value of capacitor is C = 0.62 5μF

**5.2 For Frond End Convertor integrated with LCL Filter**

For the DC Fast Charging 10 KW, using front end converter integrated with LCL filter is commonly used between the inverter and grid to reduce switching harmonics produced by grid.

Selected Value for *II* filter (LCL filter)

- Inverter side inductor  $L_1 = 5 \text{ mH}$
- Inverter side inductor  $L_2 = 5 \text{ mH}$
- Filter Capacitor  $C_f = 30 \mu \text{ F}$

Are choose manly harmonic reduce and power quality improvements, stable Current and stable grid interfacing.

**5.2.1 Modeling of DC Fast Charging Station Configuration for V2Gand G2V.**

Parameter	Value	Parameter	Value
Rated capacity	250 KW	EV rated capacity	10 KW
$V_{batt}$	360 V,50 A.	$C_{filter}$	0.625 μF
Battery Capacity	30 Ah.	$L_{filter}$	2 mH
$L_{1grid}, L_{2inv}$	5 mH, 5 mH	$C_{filter} \text{ inv side}$	30 μF

The values used in the recharging station design process are mentioned in Appendix. The wind turbine is designed in such a way that it can reach max power of 100 kW. For producing rated maximum power of 50kW, the solar PV

is simulated to operate in standard test conditions of 1000W/m<sup>2</sup> irradiance and 25° C temperature. In this system, 150-kW resistive load is linked to 480 V AC bus. For running it on unity power factor, the reactive current reference to GCI is set to Zero. EV batteries start off at around 50% state of charge. When steady-state is attained, the batteries of EV1 and EV2 as described in figure 1 help to implement V2G-G2V transfer.

### 5.2.2 Modeling simulation of DC fast charging station structure for G2V/V2G operation

#### **AFC Converter as a 3-level diode clamp multilevel inverter Modeling**

The three-level Diode-Clamped Multilevel Neutral Point Clamped (NPC) inverter serves as the front-end converter, enabling bidirectional power conversion between the DC and AC domains. During Grid-to-Vehicle (G2V) operation, the inverter converts the regulated DC output of the boost converter into three-phase AC power for grid interfacing. Conversely, during Vehicle-to-Grid (V2G) operation, it converts the three-phase AC power from the grid into DC power, which is subsequently processed by the buck converter. The proposed system employs a three-phase, three-level NPC inverter consisting of twelve MOSFET switches. The gate signals for these switches are generated using a hysteresis current controller, which continuously monitors the current error and regulates the switching states accordingly. This control strategy ensures fast dynamic response, accurate current tracking, and effective suppression of current harmonics. As a result, the output AC waveform exhibits reduced distortion and improved power quality, making the proposed converter suitable for high-performance bidirectional EV fast-charging applications.

#### **Bidirectional Buck – Boost converter Modeling**

Charging and discharging current of EV batteries is controlled with help of bidirectional DC–DC buck boost converter having two MOSFET switches, one inductor, and one capacitor by maintaining switching frequency to 10 kHz. For controlling the battery current, battery current loops are used, the said current is compared to the reference value. This generated difference passed via PI controller; PI controller's output helps in controlling the gate pulse for the MOSFET switches.

The generated reference value is positive in G2V mode, bidirectional DC–DC converter is used in buck mode to charge the battery while the generated reference value is negative in V2G mode, bidirectional DC–DC converter is used in boost mode to discharge the battery.

#### **Role of LCL filter**

The LCL filter forms the interface between the grid and the front-end converter. It minimizes the high-frequency harmonics that are introduced by the switching devices. It was also noted that the LCL filter is effective in attenuating the harmonics compared to the inductive filter while maintaining low inductance values.

The LCL filter consists of two inductors and a capacitor. The grid side and the converter side each have one inductor while one capacitor is connected across the two inductors. This configuration provides a smooth current injection to the grid.

### 5.2.3 Simulation Result and Analysis

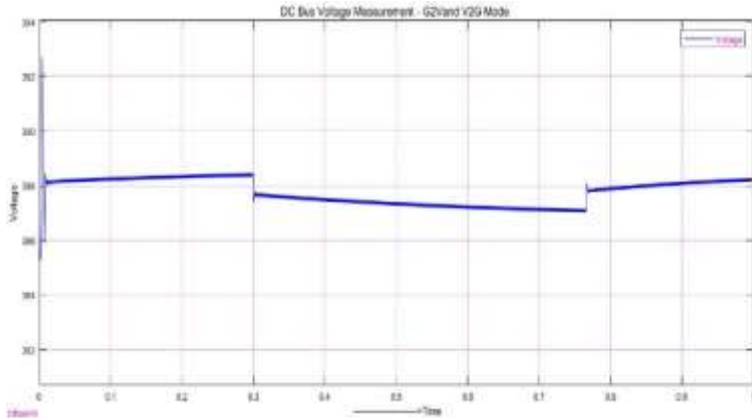
#### **DC Bus Voltage Performance**

The bus voltage level at DC is regulated at a fixed level of 800 V, as indicated in the closed-loop control system, which operates in the front-end converter.

During simulation as soon as the system gets started, the DC bus voltage immediately stabilizes at the regulated level. This indicates the significance of the 5600  $\mu$ F DC link capacitor to the stable operation of the system as a buffer between the AC grid and the DC-DC converter.

Shown in fig.6, While changing the operating mode between G2V and V2G slight change is observed in the DC-link voltage before the control system adjusts it to the reference value. This shows the effectiveness of the control tactic in the front-end converter.

#### **Fig 6 DC Bus Voltage Measurement -V2G and G2V Mode**



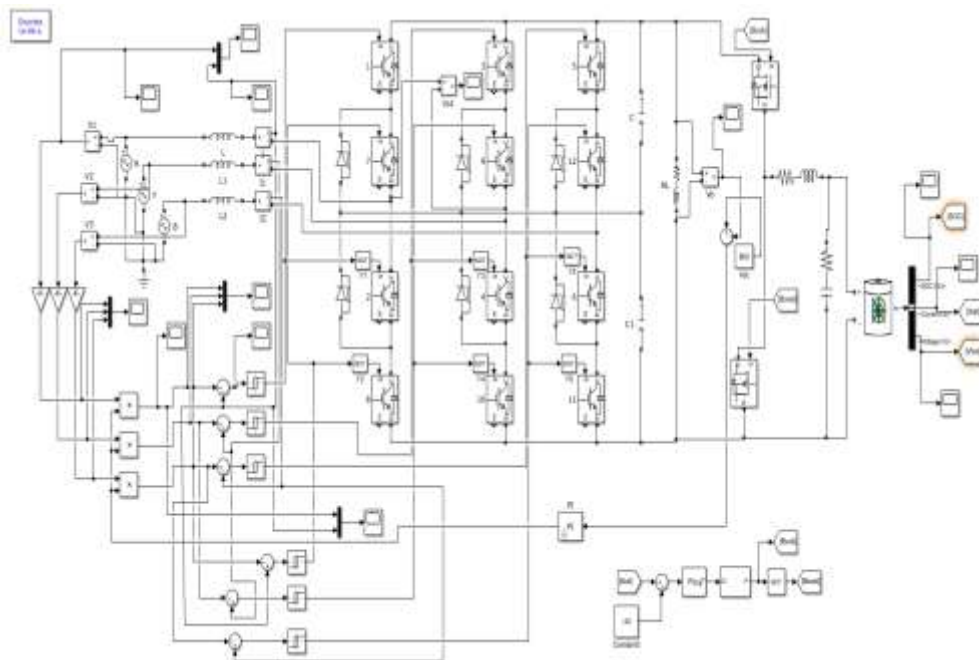
**Grid Voltage and Current Waveform.**

Shown in fig.8.the voltage and current waveforms of three-phase grid are explored to evaluate the collaborative of charging station and utility grid performance. In V2G mode, current waveform synchronizes with the waveform of grid voltage. This indicates supply of active power to the grid by Electrical Vehicles battery.

The filtering effect of the LCL filter, that successfully filters the switching harmonics produced by the inverter, ensures that the current has a sinusoidal waveform.

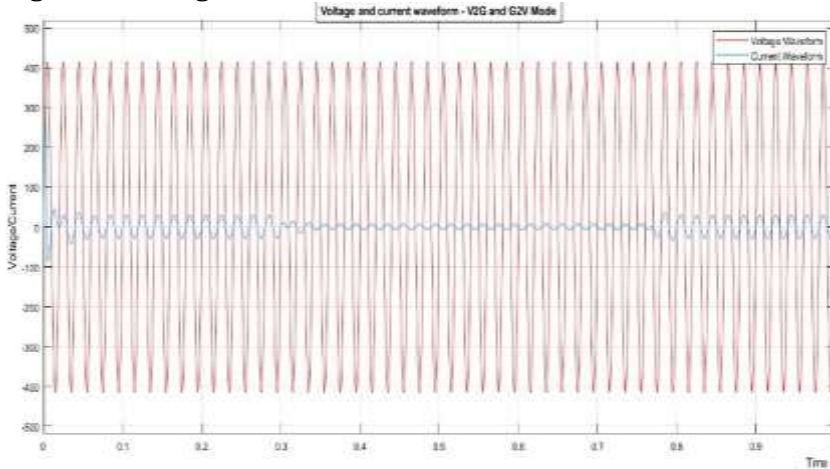
Shown in fig.9.While G2V modes on the phase relationship between voltage and currents not constant hence grid current is out of phase compared to voltage. This shows the system s receiving power from grid to charge battery. Even during phase shift the waveform quality remain stable, which indicates filters achievement for reducing harmonic distortion

**Fig 7 Modeling of DC Fast Charging Station Configuration for G2V/V2G operation**



These results show that the station for charging stays connected to the grid while running at a power factor close to one.

**Fig.8 Grid Voltage and Current Waveform – V2G and G2V Mode**



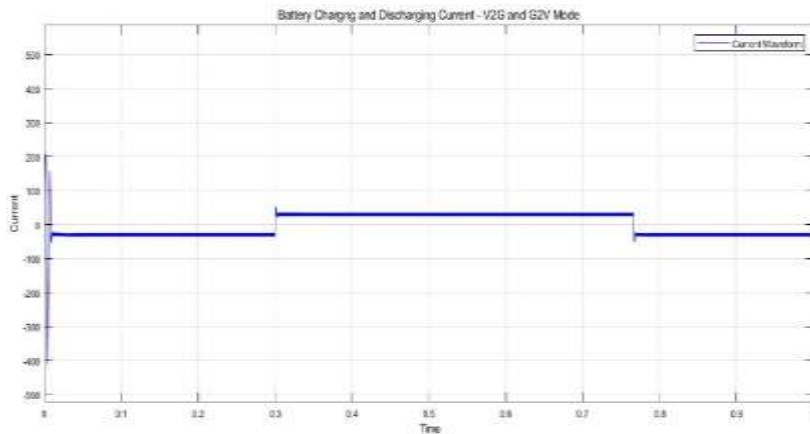
**Battery charging and discharging characteristics**

The charging station's bidirectional ability can be seen by the battery current waveform. The battery current is controlled by the current controller according to the control system's reference value.

Shown in fig.8. The battery current is positive when G2V is running, indicating that energy is moving from the electricity grid to the battery. To provide controlled battery charging, the current gradually reaches the reference charging value.

Shown in fig.9. When the battery is discharging and returning electricity to the grid, the current turns negative during V2G operation. Because a PI current controller and PWM-based switching control are used, there are not any major current spikes during the smooth change between charging and discharging modes.

**Fig 9 Battery Charging Current and Discharging Current – G2V and V2G Mode**



**Active and Reactive Power Analysis**

Shown in fig 10. The measured active power is positive during V2G operation, signifying that the EV battery is providing energy to the utility grid. As a result of this functionality, Shown n fig 11, the EV can support the grid during times of high demand by functioning as a renewable energy storage unit. G2V operation, on the other side, initiates the active power to go negative. This indicates that the energy has been obtained from the grid to charge the battery. Throughout the results of the simulation, the reactive power output stays very close to zero, indicating that the converter runs with a near-unity power factor.

Fig 10 Active Power Measurement - V2G Mode

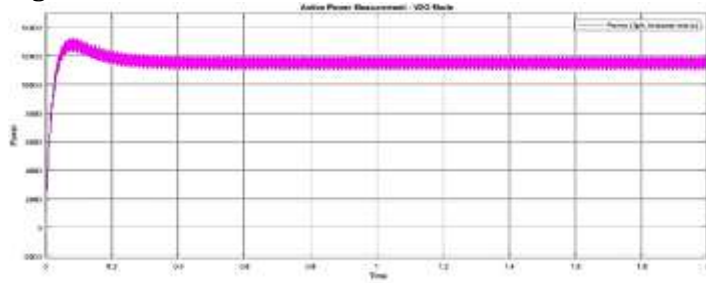
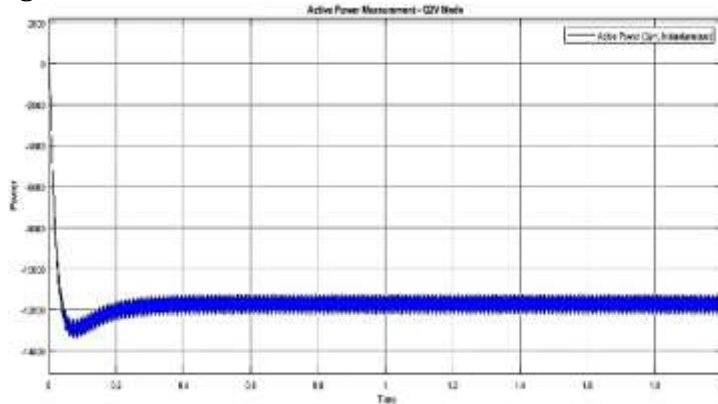


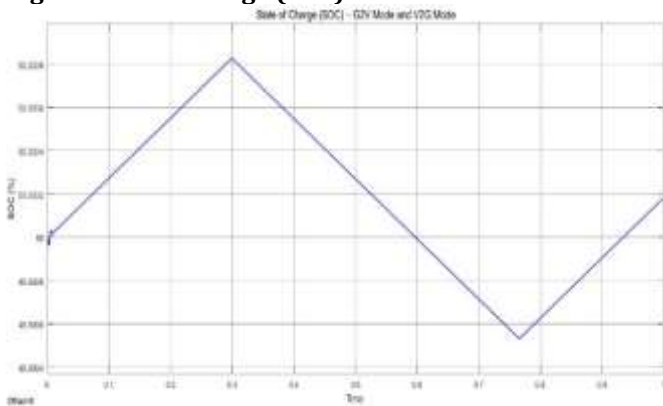
Fig 11 Active Power Measurement - G2V Mode



State of Charge (SOC) Analysis

Shown in fig 12, for the purpose to look at the energy transfer between the battery and the grid, the battery's State of Charge (SOC) is monitored during the simulation. The SOC gradually increases when electrical energy is stored in the battery during charging operation (G2V). On the flip side, as the battery provides energy to grid during V2G operation, the SOC decreases. The SOC variation exhibits a smooth profile free of sudden fluctuations, indicating that the bidirectional converter controls current flow and avoids the battery from being overloaded.

Fig 12 State of Charge (SOC) - G2V Mode and V2G Mode



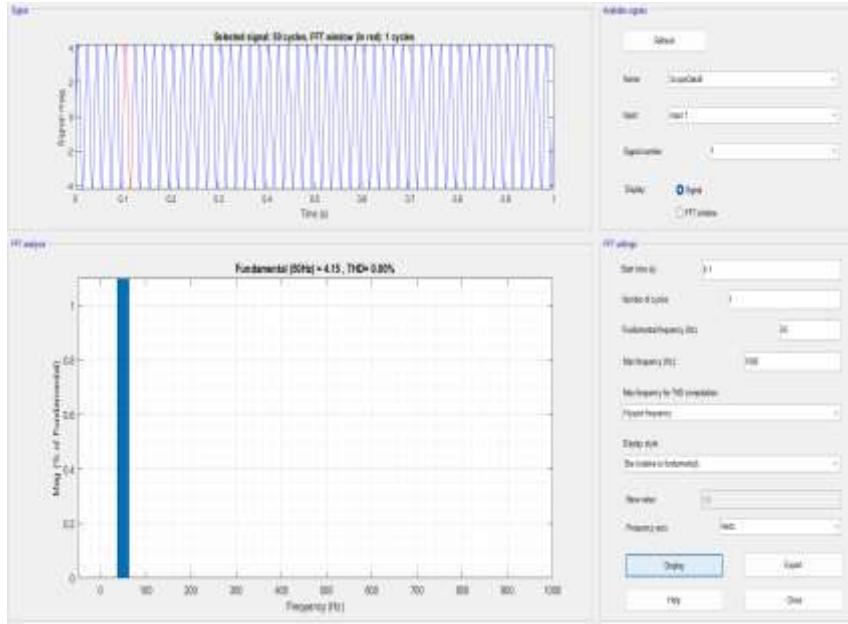
Harmonic Analysis

High-frequency harmonics generated by the DCMLI inverter's switching devices are successfully reduced by the LCL filter. As such, in comparison with systems without effective filtering, the harmonic distortion is much reduced, and the grid current maintains nearly sinusoidal. This ensures grid-connected converters adhere to power quality standards.

The control system efficiently controls battery current, maintains a steady DC bus voltage, and ensures smooth bidirectional energy transfer between the EV and the grid. The suggested system shows the viability of relating EV charging stations with the utility grid for smart energy management applications given that Hysteresis Current

control, PWM switching, and LCL filtering enable effective energy conversion while retaining good power quality. The simulation's results indicate that the suggested 10 kW bidirectional EV charging station performs effectively in both G2V and V2G modes and achieved THD result is 4.15% for injected grid Current.

**Fig 13 FFT Analysis of Grid injected current.**



**Comparative THD Analysis**

Technique	PWM Technique	THD(%)
2 Level inverter with buck boost converter – V2G and G2V Mode	SPWM	29.31%
3 Level DCMLI with buck boost converter – V2G and G2V Mode	Hysteresis Current Control	4.15%

**5. Conclusion**

This Paper is presented Modeling and design of a 10 KW V2G and G2V System in Micro Grid using DC Fast Charging architecture. A dc fast charging station with off board charger integrated with buck boost converter and front-end converter as 3 level Diode clamp Multilevel inverter to interface EVs to micro grid. The design of control system of Buck – Boost Converter and FAC will permit bidirectional power transfer among the EVs and the grid. The design and the result of simulation point towards a smooth transfer of power between the EVs, and grid and the quality of grid injected current from the EVs. In the terms of tracking the changed active power reference and dc bus voltage stability, the designed controller provides good dynamic performance. This work considers the micro grid's active power regulation elements. The suggested V2G system can be used for several different services, such as frequency regulation and reactive power control. Future study is advised to design a supervisory controller that provides command signals to the different EV charging controllers.

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